
EE273 Lecture 2 Wires

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Today's Assignment

- Reading
 - Sections 3.3.4 through 3.5.2
 - Complete before class on Wednesday

A Quick Overview

- Wires have
 - resistance
 - capacitance
 - and inductance
- To reason about wires we create models
 - ideal
 - lumped L, R, or C
 - transmission line
- Transmission lines have
 - an impedance Z_0
 - a propagation constant, A
 - from which we get velocity, v
- An LC transmission line is lossless
 - waves travel down the line without loss
$$V(x, t) = V(0, t - x/v)$$
- Waves reflect off the ends of a line depending on the termination impedance

$$V_R = V_I \left(\frac{Z_0 - R_T}{Z_0 + R_T} \right)$$

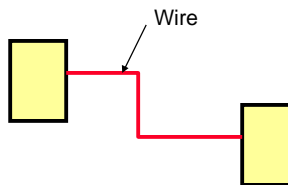
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Wires in Digital Systems

- Physically wires are
 - Stripguides on printed-circuit cards and backplanes
 - Conductors in cables and cable assemblies
 - Connectors
- We tend to treat them as **ideal wires**
 - no delay (equipotential)
 - no capacitance, inductance, or resistance
- They are **not** ideal
- To build reliable systems we need to understand their properties and **behavior**



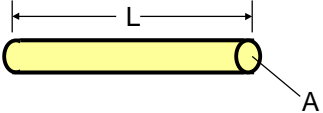
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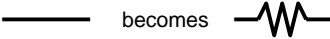
Resistance of Wires

- Most *real* wires have resistance
- Depends on
 - material (resistivity)
 - length
 - cross section
- Causes
 - delay
 - loss



$$R = \frac{\rho L}{A}$$

Material	ρ (n Ω -m)
Ag	16
Cu	17
Au	22
Al	27



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Capacitance of Wires

- Real wires have capacitance
 - line charge
 - parallel plate
 - fringing
- To compute
 - assume Q
 - compute E field
 - integrate to get V

$C = \frac{Q}{V}$

$E = \frac{Q}{2\pi r}$

$C = \frac{2\pi\epsilon}{\log\left(\frac{r_o}{r_i}\right)}$

$C = \frac{2\pi\epsilon}{\log\left(\frac{s}{r}\right)}$

$C = \frac{w\epsilon}{d} + \frac{2\pi\epsilon}{\log(2s/r)}$

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Inductance of Wires

- Real wires have inductance

$$L = \frac{\Lambda}{I}$$

- In a homogenous medium

$$CL = \epsilon\mu$$



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Some Example Wires

Type	W	R	C	L
On chip	0.6 μm	150k Ω/m	200pf/m	600nH/m
PC Board	150 μm	5 Ω/m	100pf/m	300nH/m
24AWG pair	511 μm	0.08 Ω/m	40pf/m	400nH/m

Scale model of a line has different R, but same L and C per unit length

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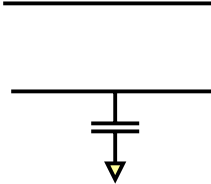
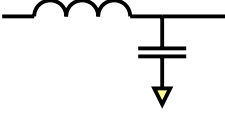
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Wire Models

- In a particular situation, we create a *model* of a wire that captures the properties we need
 - ideal
 - lumped L, R, or C
 - RC transmission line
 - LC transmission line
 - General LRCG transmission line
- Model to use depends on frequency

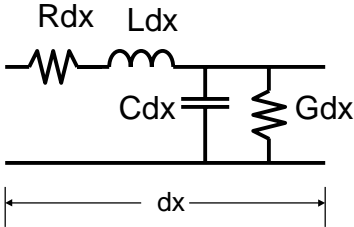
$$f_0 = \frac{R}{2\pi L}$$

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LRCG Wire Model

- Model an *infinitesimal* length of wire, dx, with lumped components
 - L, R, C, and G



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Transmission Line Equations

dx

Drop across R and L

$$\frac{\partial V}{\partial x} = RI + L \frac{\partial I}{\partial t}$$

Current into C and G

$$\frac{\partial I}{\partial x} = GV + C \frac{\partial V}{\partial t}$$

$$\frac{\partial^2 V}{\partial x^2} = RGV + (RC + LG) \frac{\partial V}{\partial t} + LC \frac{\partial^2 V}{\partial t^2}$$

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Impedance

- An infinite length of LRGC transmission line has an *impedance* Z_0
- Driving a line *terminated* into Z_0 is the same as driving Z_0
- In general Z_0 is complex and frequency dependent
- For LC lines its real and independent of frequency

=

$$Z_0 = \left(\frac{R + Ls}{G + Cs} \right)^{\frac{1}{2}}$$

$$Z_0 = \left(\frac{L}{C} \right)^{\frac{1}{2}}$$

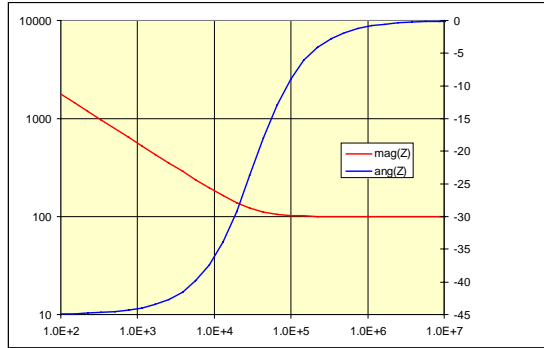
At high frequency (LC lines)

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Example, 24AWG Pair

- $f_0 = 33\text{kHz}$
- Below f_0 , line is RC
- Above f_0 , line is LC

$$Z_0 = \left(\frac{.08 + 400 \times 10^{-9} \times 2\pi f j}{40 \times 10^{-9} \times 2\pi f j} \right)^{\frac{1}{2}}$$



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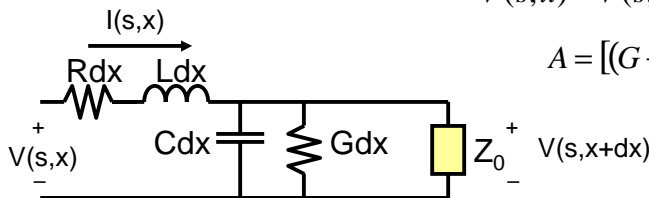
Propagation Constant

- Using impedance, we can solve for $V(s,x)$
- Propagation is governed by a constant, A
 - real part is attenuation
 - imaginary part is phase shift
 - velocity⁻¹

$$\begin{aligned} \frac{\partial V(s)}{\partial x} &= -(R + Ls)I(s) \\ &= -(R + Ls)V(s)/Z_0 \\ &= -[(G + Cs)(R + Ls)]^{\frac{1}{2}} V(s) \end{aligned}$$

$$V(s, x) = V(s, 0) \exp(-Ax)$$

$$A = [(G + Cs)(R + Ls)]^{\frac{1}{2}}$$



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Lossless LC Lines

- If R and G are negligible
 - line is lossless (no dissipation)
 - governed by the *wave equation*
- Waves propagate down the line in both directions without distortion
- Line is described by its impedance and velocity
- What happens when the wave gets to the end of the line?

$$\frac{\partial^2 V}{\partial x^2} = LC \frac{\partial^2 V}{\partial t^2}$$

$$V_f(x, t) = V \left(0, t - \frac{x}{v} \right)$$

$$V_r(x, t) = V \left(x_{\max}, t - \frac{x_{\max} - x}{v} \right)$$

$$v = (LC)^{-\frac{1}{2}}$$

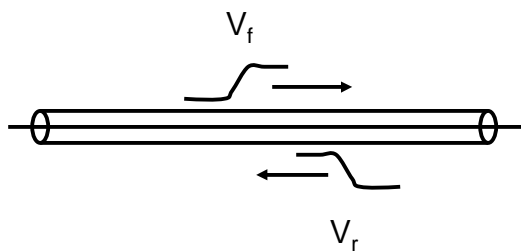
$$Z_0 = \left(\frac{L}{C} \right)^{\frac{1}{2}}$$

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Lossless LC Line



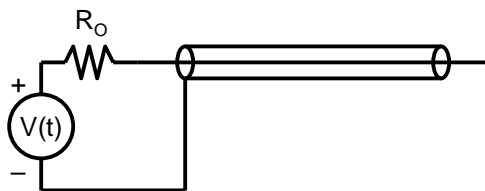
Waveform on line is superposition of forward and reverse traveling waves

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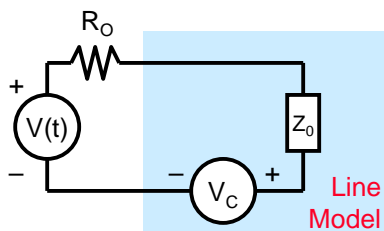
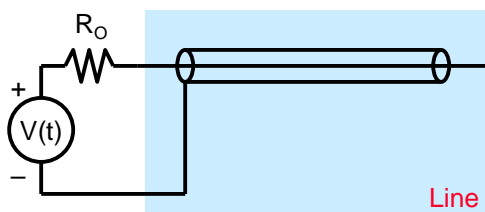
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Driving a Transmission Line



Place waves on the line by driving one end with a source
Assume line is infinite for now

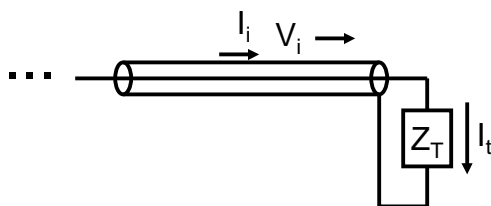
Driving a Line - Equivalent Circuit



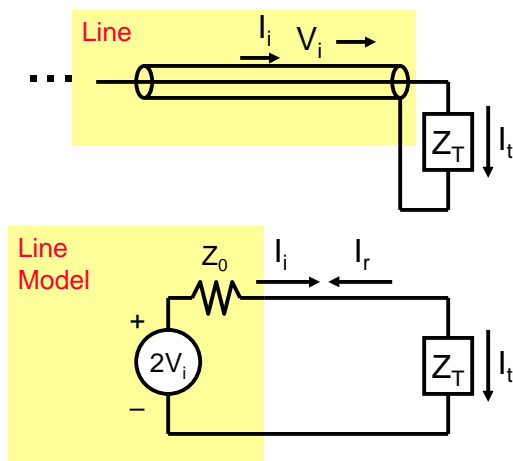
Response of line to voltage source depends on previous state of line, V_c

Termination

- Suppose we drive a unit step, $U(t)$, on the line
- What happens at the far end?



Termination - Equivalent Circuit



Reflections and The Telegrapher's Equation

- Incident wave determines V_i, I_i
- Use equivalent circuit to solve for V_T, I_T
- Use superposition to calculate V_r, I_r

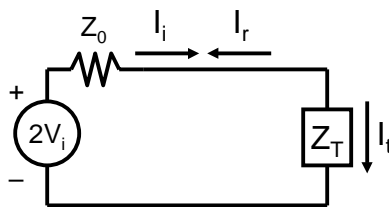
$$I_T = \frac{2V_i}{Z_0 + Z_T}$$

$$I_r = I_i - I_T$$

$$I_r = \frac{V_i}{Z_0} - \frac{2V_i}{Z_0 + Z_T}$$

$$I_r = \frac{V_i}{Z_0} \left(\frac{Z_T - Z_0}{Z_T + Z_0} \right)$$

$$\frac{I_r}{I_i} = \frac{V_r}{V_i} = \left(\frac{Z_T - Z_0}{Z_T + Z_0} \right)$$

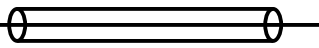


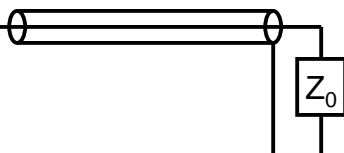
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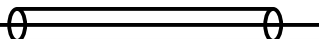
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Some Common Terminations

...  Open circuit

...  Matched termination

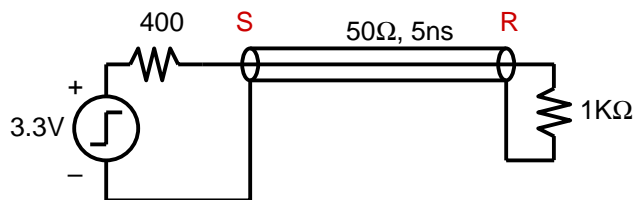
...  Short circuit

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Example of Reflections

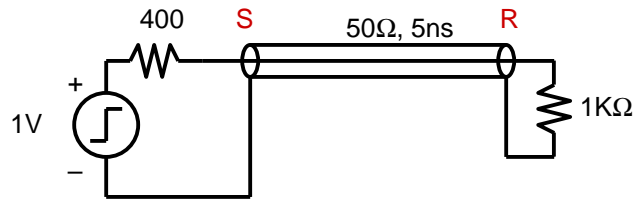


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Example of Reflections



$$V_i = 1V \left(\frac{50}{400 + 50} \right) = 0.111V$$

$$k_{rR} = \frac{1000 - 50}{1000 + 50} = 0.905$$

$$k_{rS} = \frac{400 - 50}{400 + 50} = 0.778$$

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Example of Reflections

1V source, 400 Ω resistor, switch S, 50 Ω , 5 ns transmission line, switch R, 1K Ω load resistor.

$$k_{rR} = \frac{1000 - 50}{1000 + 50} = 0.905$$

$$V_i = 1V \left(\frac{50}{400 + 50} \right) = 0.111V$$

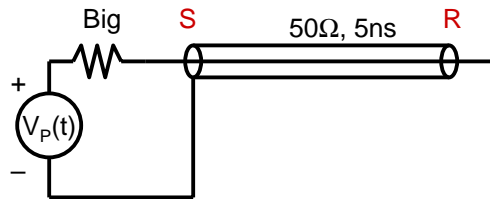
$$k_{rS} = \frac{400 - 50}{400 + 50} = 0.778$$

	Vwave	Vline	t
Vi1	0.111	0.111	0
Vr1	0.101	0.212	5
Vi2	0.078	0.290	10
Vr2	0.071	0.361	15
Vi3	0.055	0.416	20
Vr3	0.050	0.465	25
Vi4	0.039	0.504	30
Vr4	0.035	0.539	35
Vi5	0.027	0.566	40

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Standing Waves



What happens if we drive an open line with a sine wave?
With an arbitrary periodic function (period = round trip)
What if the line is shorted?

Next Time

- Lossy wires
 - attenuation
 - RC transmission lines
- Special transmission lines
 - multi-drop buses
 - balanced lines
 - even and odd-mode propagation